

2. Signal Type and Analysis Method

- 2.1 Stationery Signal analysis Methods
 - 2.1.1 Statistics Method
 - (1) Gaussian Distribution and parameters
 - (2) Evaluation of different from Gaussian distribution
 - 2.1.2 Statistic Model
 - 2.1.3. Spectrum Analysis
 - (1)Fourier Transform
 - (2) Maximum Entropy Method (MEM)
- 2.2 Non-stationary Signal Analysis Method
 - 2.2.1 Short Time Fourier transform(STFT)
 - 2.2.2 Continuance Wavelet Transform (CWT)
- 2.3 Example: sound source separation

Sound source separated By ICA and CDWT

Important point of last lesson

1) Signal Type and Analysis Method

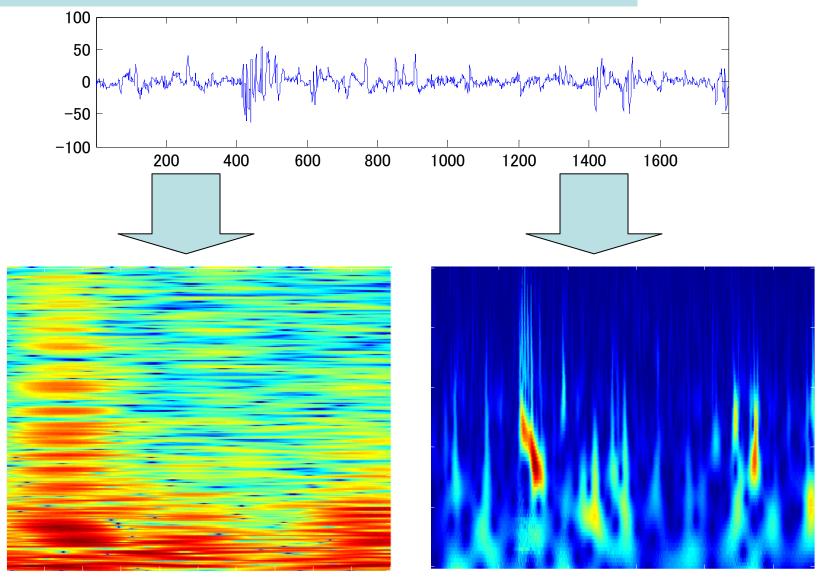
Signal Type and Comparing Analysis Methods

Wave Method	Statistics Method	Statistics Model	Spectrum Analysis (FFT, MEM)	STFT, WT
Sine wave	0	0	©	0
Compound cycle	0	0	0	0
Steady (Ergodic)	0	0	0	0
non-Ergodicity	×	×	×	©
Unsteady	×	×	×	0

*Method adapted for unsteady signal analysis:

- 1) Short Time Fourier transform, STFT
- 2) Wavelet Transform, WT

2) Characteristics of the CWT and STFT



CWT (Continuum Wavelet Transform)

3.1 Introducing Wavelet Transform

3.1.1 Continuance Wavelet Transform (DWT)

1) Definition of the CWT

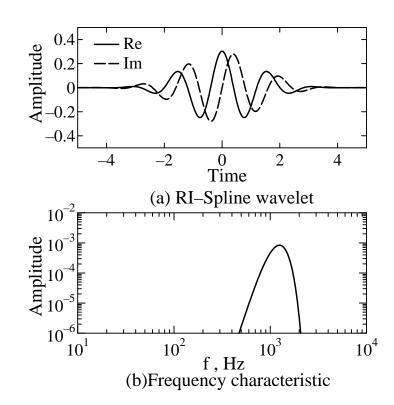
$$w(a,b) = \int_{-\infty}^{\infty} f(t) \overline{\psi}_{a,b}(t) dt$$

$$\psi_{a,b}(t) = \frac{1}{\sqrt{a}} \psi(\frac{t-b}{a})$$

 $\psi(t)$: Mother Wavelet (MW)

a: Scale (1/a Frequency)

b: Time



2) Features of the CWT

Advantages:

- Transforming signal to time-frequency plane and making its nature clearly (illustrating)
- Time and frequency resolution can be changed with frequency

Problems:

- Redundancy bases
- How do select Mother wavelet?
- There isn't a common fast algorithm for calculation.

3000

(b) Basis of CWT for 6 octave, 4 divided

4000

3) Example of the CWT

$$w(a,b) = \int_{-\infty}^{\infty} f(t) \overline{\psi}_{a,b}(t) dt$$
$$\psi_{a,b}(t) = \frac{1}{\sqrt{a}} \psi(\frac{t-b}{a})$$

 $\psi(t)$: Mother Wavelet (MW)

a: Scale (1/a Frequency)

b: Time

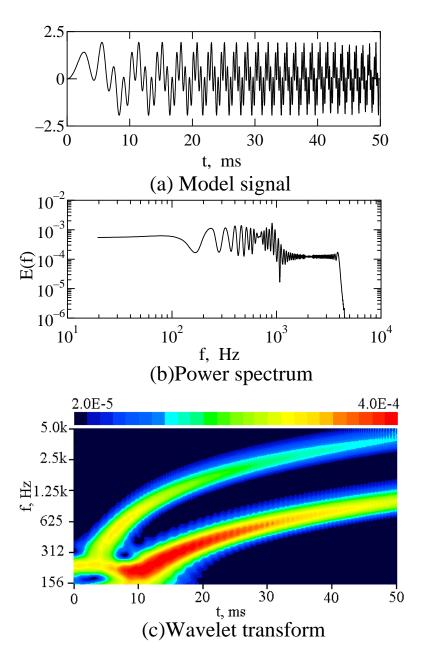


Fig.1 Continuance Wavelet Transform of Model Signal

3.1.2 Discrete Wavelet Transform (DWT)

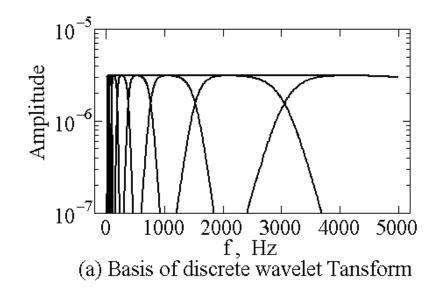
1)Definition of the DWT

$$d_k^j = \int_{-\infty}^{\infty} f(t) \overline{\psi}_{j,k}(t) dt$$

$$\psi_{j,k}(t) = 2^{j/2} \psi(2^j t - k)$$

$$a = 2^j$$

$$b = k2^j$$



It is octave analysis so bases are not overlap (Non-redundancy)

2)Fast algorithm Based on MRA (by Mallat)

$$d_{k}^{j-1} = \sum_{k} a_{l-2k} c_{k}^{j}$$

$$c_{k}^{j-1} = \sum_{k} b_{l-2k} c_{k}^{j}$$

$$c_{k}^{0} \xrightarrow{d_{k}^{-1}} c_{k}^{-2} \xrightarrow{d_{k}^{-M}} \cdots \xrightarrow{d_{k}^{-M}} c_{k}^{-M}$$
(a) Decomposition Tree
$$c_{k}^{0} \xrightarrow{d_{k}^{-1}} c_{k}^{-M} \xrightarrow{d_{k}^{-M+1}} d_{k}^{-M}$$
(b) Reconstruction Tree

Fig.2 Tree algorithm of MRA

3) Features of the (Real) DWT



Advantages:

- •Good compression of signal energy.
- Perfect reconstruction with short support filters.
- •Non-redundancy. Very low computation cost, order-N only.

Problems:

- Severe shift dependence.
- •Poor directional selectivity in 2-D,3-D etc.

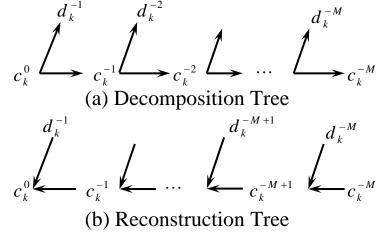


Fig.2 Tree algorithm of MRA

The DWT is normally implemented with a tree of high pass and low pass filters, which proposed by Mallat (fast algorithm).

4) Example of the DWT

$$d_k^j = \int_{-\infty}^{\infty} f(t) \overline{\psi}_{j,k}(t) dt$$

$$\psi_{j,k}(t) = 2^{-j/2} \psi(2^j t - k)$$

$$j: \text{Level}$$

k: Time

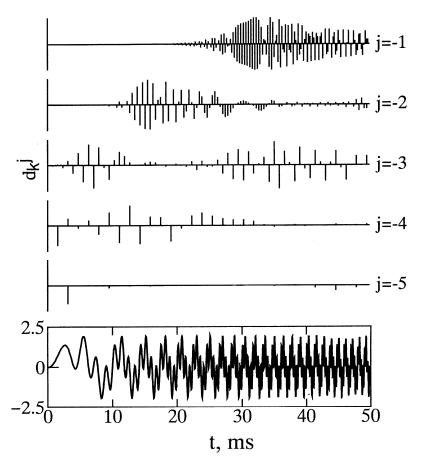


Fig.3 Discrete Wavelet Transform of Model Signal

3.2 Selection of MW for the CWT

3.2.1 Condition of MW selection

$$w(a,b) = \int_{-\infty}^{\infty} f(t)\overline{\psi}_{a,b}(t)dt \qquad \begin{array}{c} \psi(t) : \text{Mother Wavelet (MW)} \\ a : \text{Scale (1/a Frequency)} \end{array}$$
$$\psi_{a,b}(t) = \frac{1}{\sqrt{a}}\psi(\frac{t-b}{a}) \qquad b : \text{Time}$$

Condition of The MW (admissibility condition):

$$C_{\psi} = \int_{-\infty}^{\infty} \frac{|\hat{\psi}(\omega)|^2}{|\omega|} d\omega < \infty$$

This condition can be sampled as following Equation:

$$\int_{-\infty}^{\infty} \psi(t)dt = 0$$

So many MW have been proposed and how do select them be comes problem.

Example of the MW

Mother Wavelet	Composition	Charactreistic	
Gauss type	$\psi(t) = exp(\frac{-t^2}{2})[exp(i\Omega t) - exp(-\frac{\Omega^2}{2})]$ $\Omega = 2\pi\xi$	- Real Imaginary $\Delta t \Delta \omega = 0.5001$	
Gabor	$\psi(t) = \pi^{-1/4} (\omega^*/\gamma)^{1/2} exp[-(t\omega^*/\gamma)^2/2 - i\omega^* t]$ $\omega^* = 2\pi f^*$ $\gamma = \pi (2/\ln 2)^{1/2} \approx 5.336$	- Real Imaginary $\Delta t \Delta \omega = 0.5002$	
Daubechise(6)	$\psi(t) = \sum_{k=0}^{2N-1} (-1)^k \bar{p}_{-k+1} \phi(2t-k)$ $\begin{cases} \phi(t) = \lim_{n \to \infty} \phi_n(t) \\ \phi_n(t) = \sum_{k=0}^{N} p_k \phi_{n-1}(t-k), n = 1, 2, \cdots \\ \phi_0(t) = N_2(t) \end{cases}$	$\Delta t \Delta \omega = 0.9055$	
Meyer	$H(\omega) = \begin{cases} exp(1/\omega^2), & \omega > 0 \\ 0, & \omega \le 0 \end{cases}$ $G(\omega) = H(4\pi/3 - \omega)/[H(\omega - 2\omega/3) + H(4\pi/3 - \omega)]$ $\phi(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} G(\omega)e^{i\omega t}d\omega$ $\psi(t) = 2\phi(2t - 1) - \psi(t - 1/2)$	$\Delta t \Delta \omega = 0.6040$	
Spline(5)	$\begin{split} N_m(t) &= \frac{t}{m-1} N_{m-1}(t) + \frac{m-t}{m-1} N_{m-1}(t-1), t \in \mathbf{R} \\ N_{2m}(k) &= \frac{k}{2m-1} N_{2m-1}(k) + \frac{2m-k}{2m-1} N_{2m-1}(k), k \in \mathbf{Z} \\ q_n &= \frac{(-1)^n}{2^{m-1}} \sum_{l=0}^m C_l^m N_{2m}(n+1-l), n = 0, \cdots, 3m-2 \\ \psi(t) &= \sum q_n N_m(2t-n), n = 0, \cdots, 3m-2 \end{split}$	$\Delta t \Delta \omega = 0.5010$	

Analysis results (1)

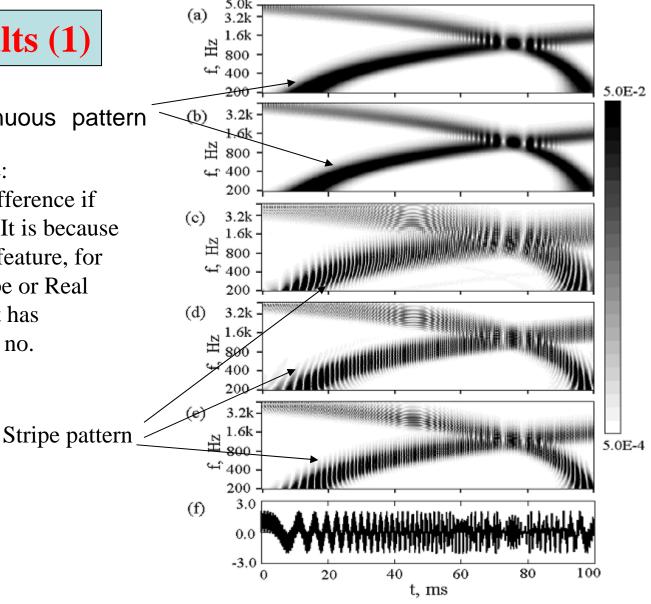
Continuous pattern

As is sown in example: Analysis results are difference if we use different MW. It is because

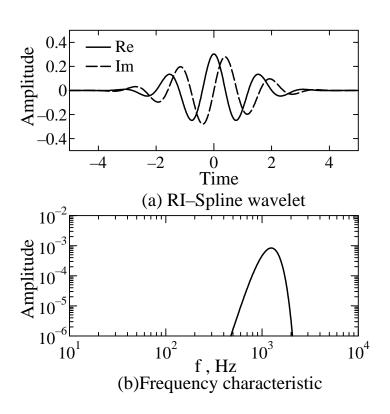
the MW has different feature, for

example, Complex type or Real Type. For Real type, it has

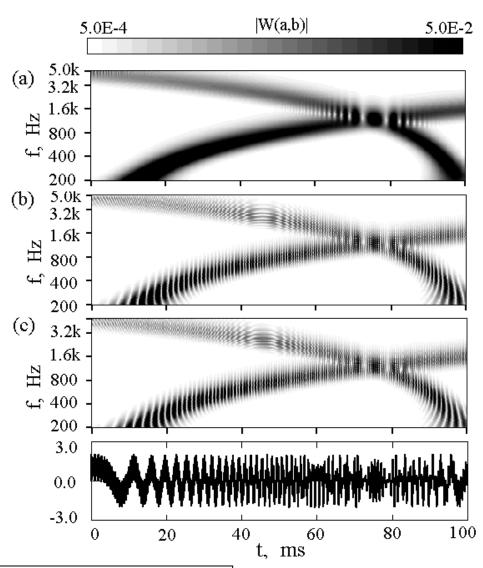
symmetric property or no.



Analysis results (2)



What is reason of non-continuance pattern?

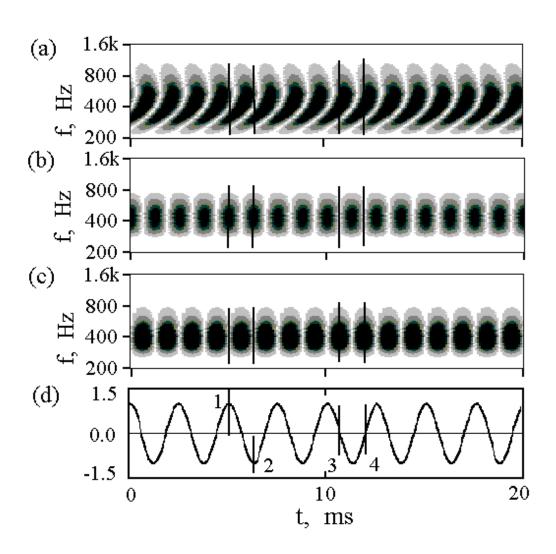


Real MWs Phase (symmetric)

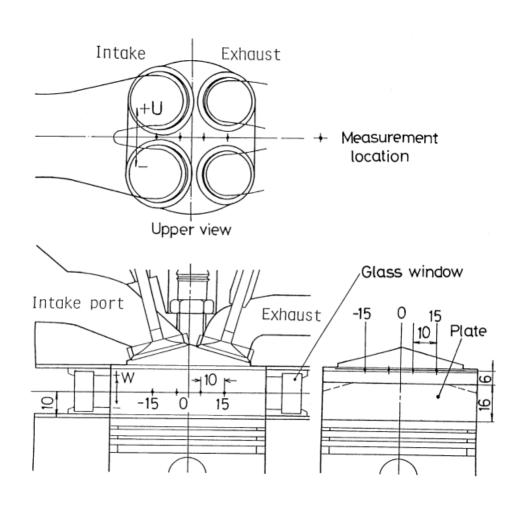
Analysis results (2)

Difference results can be obtained from different MW which has different symmetric property

How do select the MW becomes Problem, although MW is chosen freely.



3.2.2 Example of flow turbulence analysis



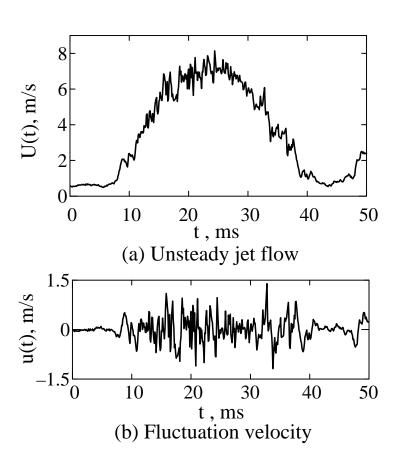
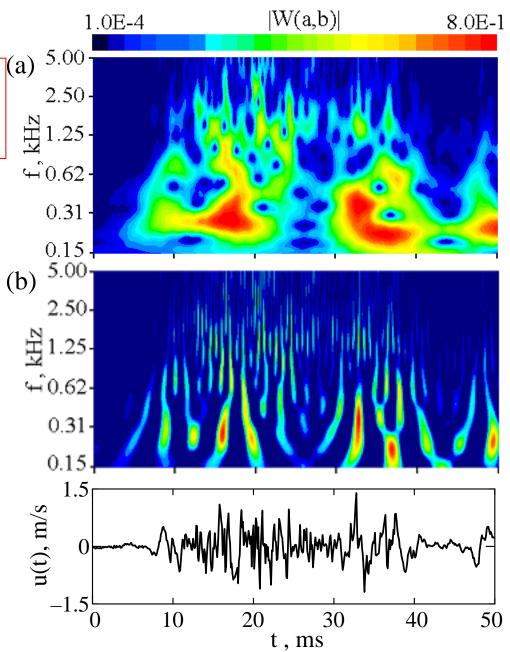


Fig.8 Example of unsteady jet flow U(t) and its fluctuation velocity u(t)

Result obtained by CWT Using different MW

Fig. Results obtained by CWT, where (a) is RI-Spline wavelet, (b) is m=4 Spline wavelet. Signal is turbulent flow with swirl in the cylinder of the engine.



3.3 Fast Algorithm for the CWT

3.3.1 Fast Algorithm in frequency domain

$$w(a,b) = \int_{-\infty}^{\infty} f(t)\overline{\psi}_{a,b}(t)dt$$
$$= \frac{1}{\sqrt{a}} \int_{-\infty}^{\infty} f(t)\overline{\psi}(\frac{t-b}{a})dt$$

It is convolution integral, so it can be changed to next equation:

$$w(a,b) = a^{1/2} \int_{-\infty}^{\infty} X(f) \overline{\psi}(af) e^{i2\pi fb} df$$

By using this equation, CWT can be did fast.



Comparing calculation cast

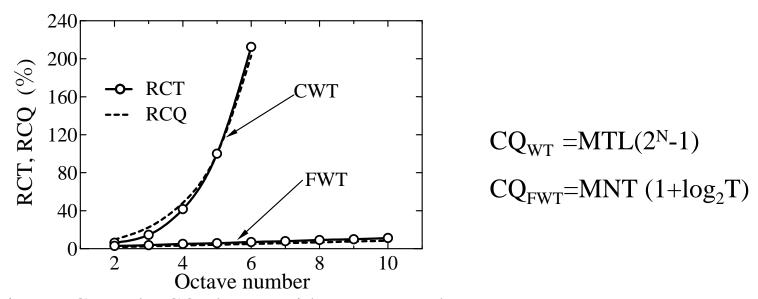
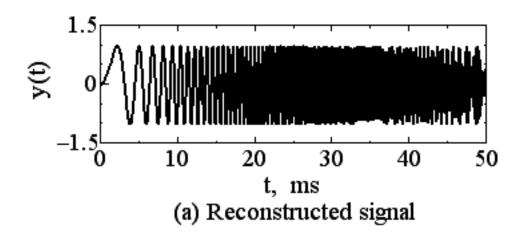


Fig. 9 RCT and RCQ change with octave number

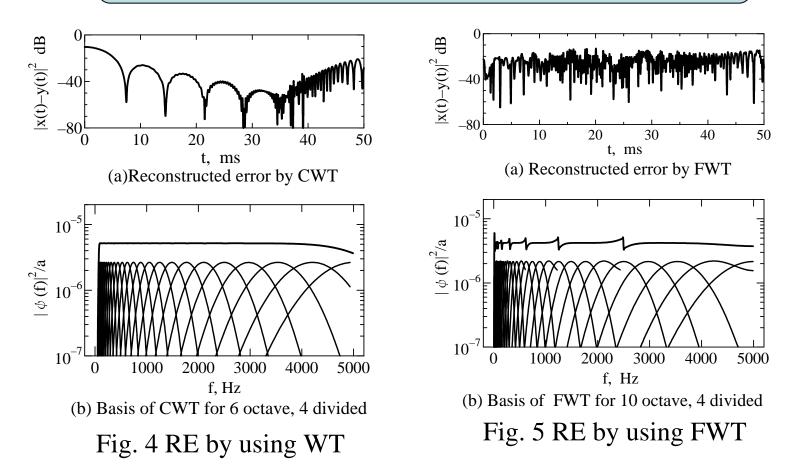
Fig. 9 shows RCT and RCQ (Ratio of Computation Quantity) which change with the increase in the number of analysis octaves, and Both RCT and RCQ are expressed with the ratio that setting the value of WT in five octaves as 100.

For comparing calculation accuracy, next signal has been used. The feature of the signal is that its frequency change with time, so it is suitable for test signal analysis method.



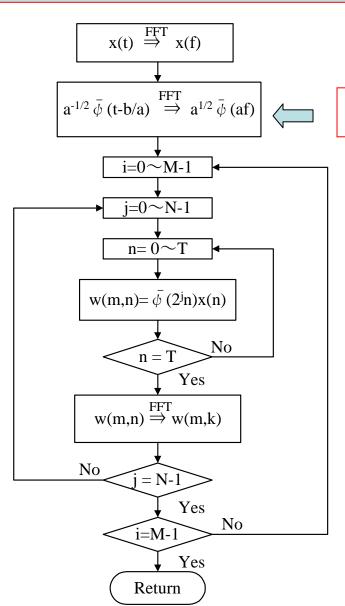


Comparing with conditional algorithm



But calculation accuracy is not good by comparing with conditional method.

3.3.2 Improvement Fast Algorithm



Improving calculation accuracy

Reason: number of MW is small. For improving calculation accuracy, we need using long MW with large number.

- 1) The length of data is doubled four and FFT of MW is performed.
- 2) It uses taking out one fourth of the obtained data.



Improving Calculation Accuracy

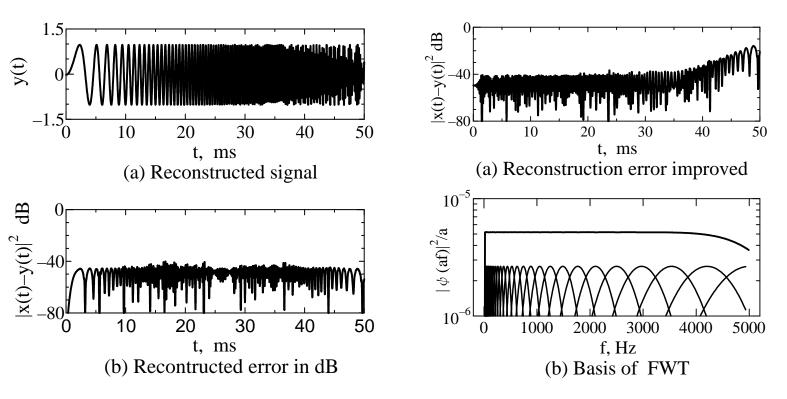


Figure 7 RE by using FWTH

Fig. 6 RE by using FWT based CBFA

Best calculation accuracy has be obtained



3.3.3 Example of EEG analysis



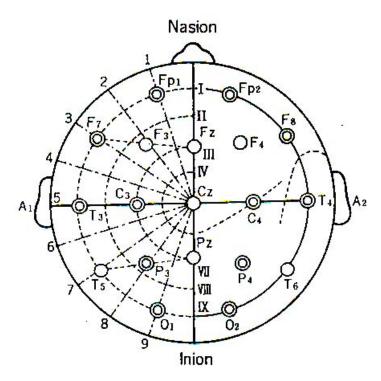


Fig. 10 Measured points of EEG by use Ten-twenty electrode system

Example of EEG waves

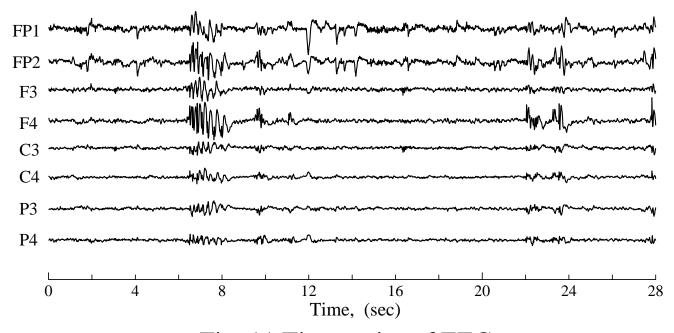
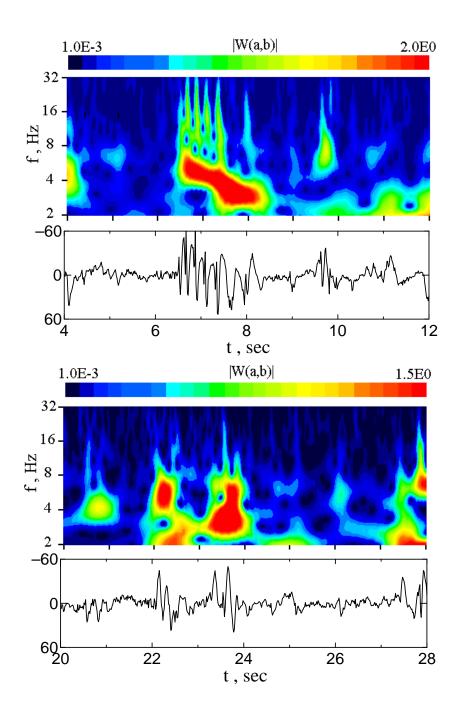


Fig. 11 Time series of EEG

Figure shows EEGs of a 14-year-old girl recorded when she was sitting in a chair and opening her eyes. The data sampling frequency is 64Hz. The spike and wave complex (SWC) around 3Hz are recorded near 6-8, and 22-24 sec. It is strong in the incipit and its generation source is in the depths.



EEG Analysis

Defined waves in EEG as:

 δ : 2-4 Hz

 θ : 4-8 Hz

 α : 8-13 Hz

β: 13-30Hz

Computation time:

FWT is only 17% of CWT

Figure shows wavelet transform of FP2 using the FWT, where the ordinates denote frequency, transverse time and the amplitude |w(a,b)| is shown as the color label. Frequency range was chosen as four octaves and each octave was divided into 48 voices for clarity.



δ , θ , α , β Waves Change with Tim

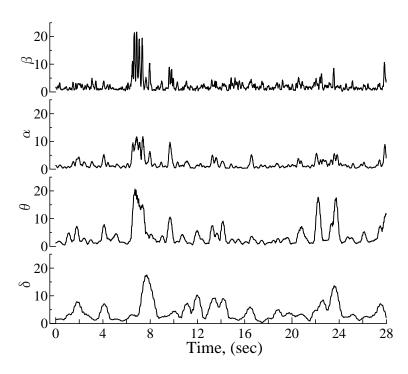


Fig. 13 δ , θ , α , β waves of EEG in point FP2

As is shown in Fig.13, we can observe the active change of the δ , θ , α , β waves before and after SWC occurred between 6-8 sec. The same phenomenon was observed before and after SWC occurred between 22-24 sec.



Comparing Calculation Accuracy

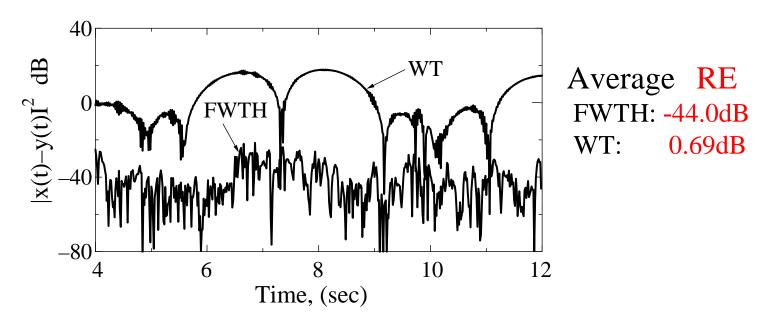


Fig.14 RE of EEG in point FP2 from 4 to 12 sec by using FWTH and WT As shown in Fig. 14, the accuracy obtained with FWTH is higher than WT. The average value of the RE of FWTH between 4-12 sec is -44dB and WT is 0.69dB. Therefore, we may conclude that our approach proposed in this study is effective for EEG analysis in real time with high accuracy and is furthermore useful for general signal processing.

3.4 Constructing new RMW

Original Signal s(t)

- (1) Hanning Window
 - (2) Normalize
- (3) Fourier Transform $rmw(t) \rightarrow \widehat{rmw}(\omega)$

(4) Hilbert Transform

$$\hat{Z}(\omega) = \begin{bmatrix} 2\widehat{rmw}(\omega) & \omega > 0 \\ \widehat{rmw}(\omega) & \omega = 0 \\ 0 & \omega < 0 \end{bmatrix}$$



$$\hat{Z}(\omega) = egin{pmatrix} \sqrt{\hat{Z}(\omega)_{real}^2 + \hat{Z}(\omega)_{imaginary}^2} & \text{real} \\ 0 & \text{imaginary} \end{pmatrix}$$

(6) Inverse Fourier Transform $SCRMW(t) \leftarrow \hat{Z}(\omega)$

Symmetric Complex RMW SCRMW(t)

Real signal Mother Wavelet

Condition
$$C_{\psi} = \int_{-\infty}^{\infty} \frac{|\hat{\psi}(\omega)|^2}{|\omega|} d\omega < \infty$$
 $\int_{-\infty}^{\infty} \psi(t) dt = 0$

$$\|\psi_R\| = 1$$

Complex RMW

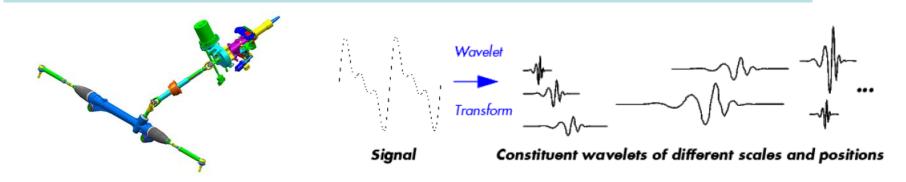
Phase imformation

Symmetric Complex RMW

It's suitable for unsteady signal

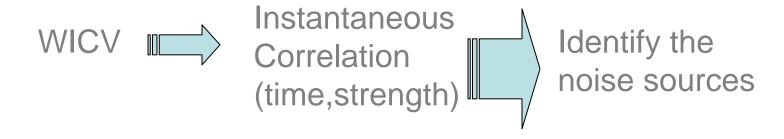


Correlation between Noise and Vibration



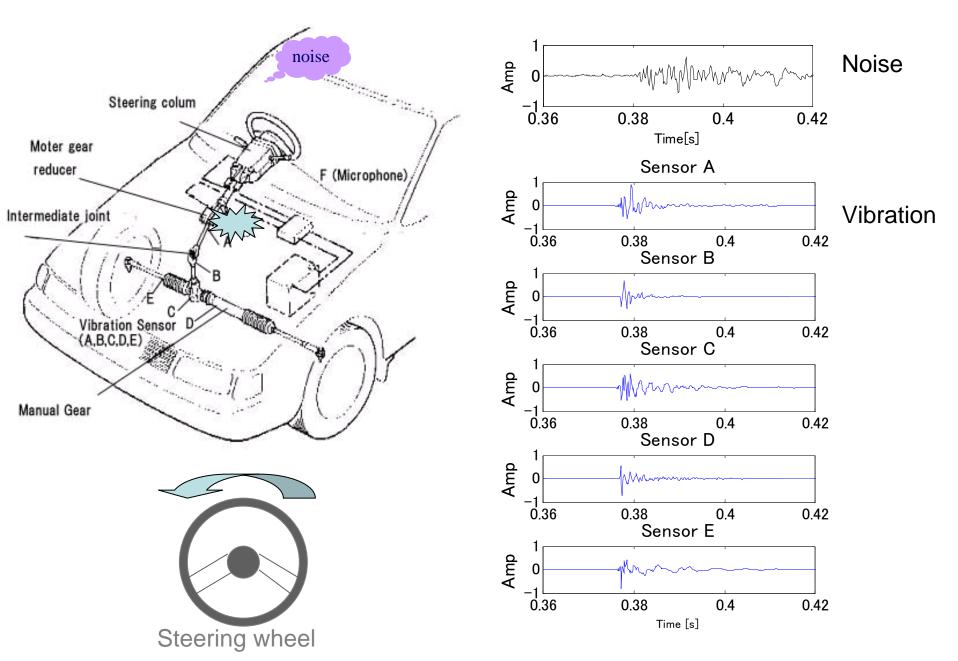
Wavelet Instantanious Correlation Value (WICV)

$$R(b) = \left| W_{(a=1,b)} \right| = \int_{-\infty}^{\infty} f(t) \psi_R(t) dt$$



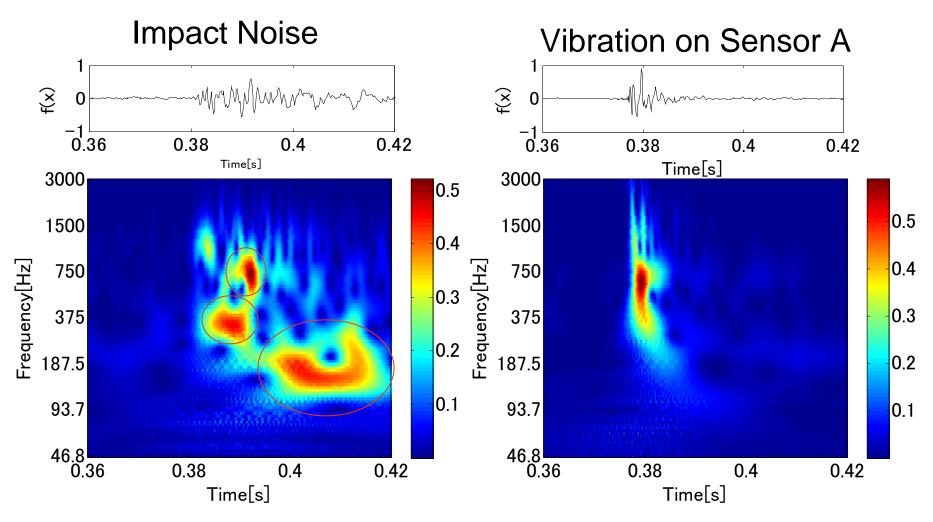


Application on Impact Noise in Case 1





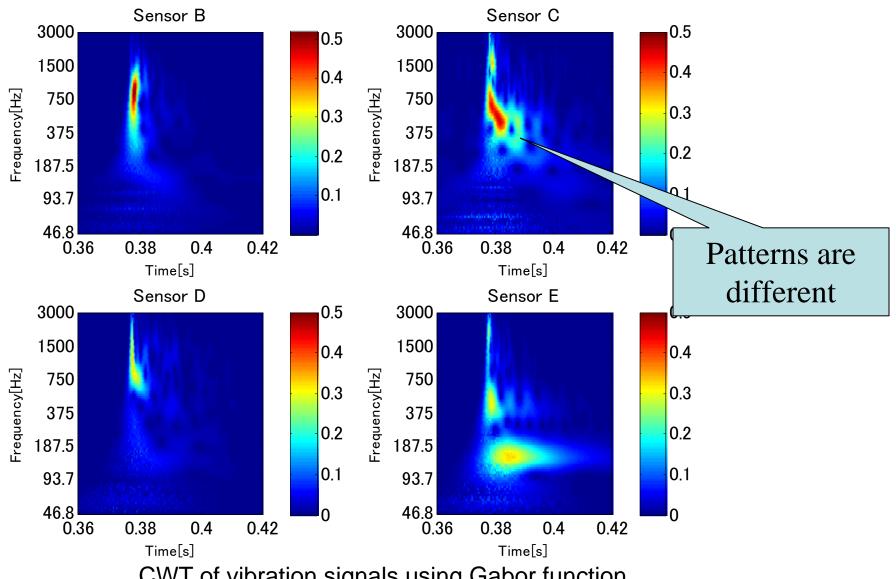
Example of CWT



CWT of noise using Gabor function

CWT of vibration using Gabor function

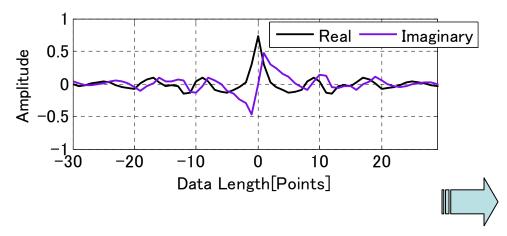
Example of CWT



CWT of vibration signals using Gabor function

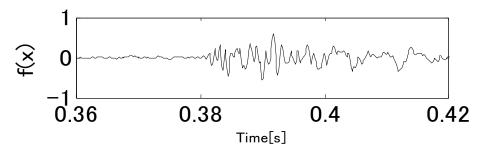


Wavelet Instantaneous Correlation (WIC)



Scale 0:same frequency charactaristic

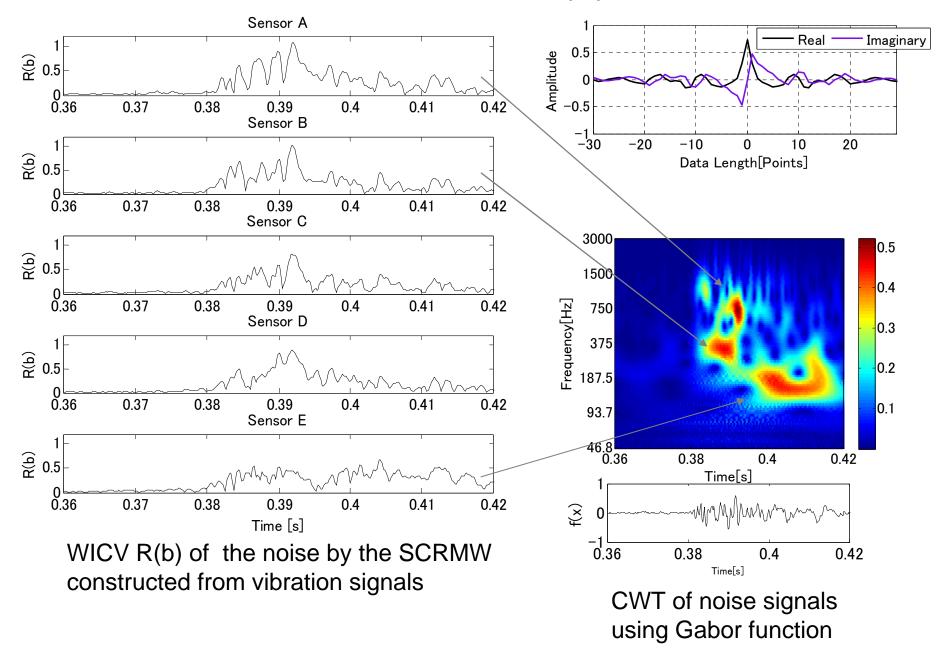
$$R(b) = \left| W_{(a=1,b)} \right| = \int_{-\infty}^{\infty} f(t) \psi_R(t) dt$$



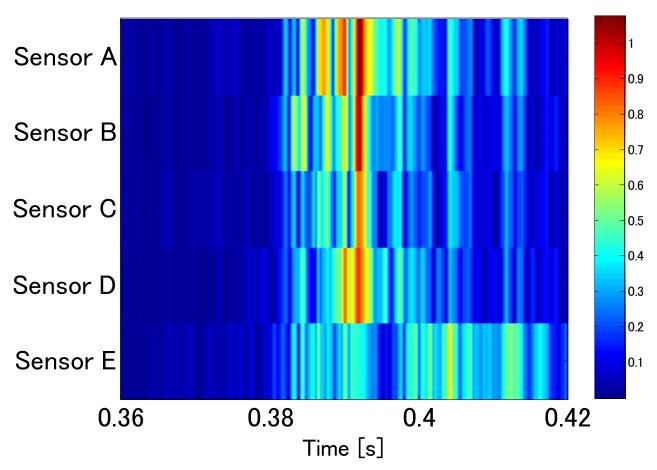
WIC using SCRMW constructed the vibration



Results of WICV R(b)



Results of WICV R(b)



WICV R(b) of the noise by the SCRMW constructed from vibration signals

3. Contiunance Wavelet Transform

- 3.1 Introducing Wavelet Transform
 - 3.1.1 Continuance Wavelet Transform (DWT)
 - 3.1.2 Discrete Wavelet Transform (DWT)
- 3.2 Selection of MW for the CWT
 - 3.2.1 Condition of MW selection
 - 3.2.2 Example of flow turbulence analysis
- 3.3 Fast Algorithm for the CWT
 - 3.3.1 Fast Algorithm in frequency domain
 - 3.3.2 Improvement Fast Algorithm
 - 3.3.3 Example of EEG analysis
- 3.4 Constructing new RMW

Wavelet Instantaneous Correlation (WIC)